## Features

- Independently Drives 4 N-Channel FET in Half Bridge or Full Bridge Configurations
- Bootstrap Supply Max Voltage to $95 \mathrm{~V}_{\text {DC }}$
- Drives 1000 pF Load at 1 MHz in Free Air at $50^{\circ} \mathrm{C}$ with Rise and Fall Times of Typically 10ns
- User-Programmable Dead Time
- On-Chip Charge-Pump and Bootstrap Upper Bias Supplies
- DIS (Disable) Overrides Input Control
- Input Logic Thresholds Compatible with 5V to 15V Logic Levels
- Very Low Power Consumption


## Applications

- Medium/Large Voice Coil Motors
- Full Bridge Power Supplies
- Class D Audio Power Amplifiers
- High Performance Motor Controls
- Noise Cancellation Systems
- Battery Powered Vehicles
- Peripherals
- U.P.S.


## Ordering Information

| PART <br> NUMBER | TEMP. <br> RANGE $\left({ }^{\circ} \mathbf{C}\right)$ | PACKAGE | PKG. <br> NUMBER |
| :---: | :---: | :---: | :---: |
| HIP4081IP | -40 to 85 | 20 Lead Plastic DIP | E20.3 |
| HIP4081IB | -40 to 85 | 20 Lead Plastic SOIC | M20.3 |

## Description

The HIP4081 is a high frequency, medium voltage Full Bridge N-Channel FET driver IC, available in 20 lead plastic SOIC and DIP packages. The HIP4081 can drive every possible switch combination except those which would cause a shoot-through condition. The HIP4081 can switch at frequencies up to 1 MHz andssssss is well suited to driving Voice Coil Motors, high-frequency Class D audio amplifiers, and power supplies.
For example, the HIP4081 can drive medium voltage brush motors, and two HIP4081s can be used to drive high performance stepper motors, since the short minimum "on-time" can provide fine micro-stepping capability.
Short propagation delays of approximately 55 ns maximizes control loop crossover frequencies and dead-times which can be adjusted to near zero to minimize distortion, resulting in rapid, precise control of the driven load.
A similar part, the HIP4080, includes an on-chip input comparator to create a PWM signal from an external triangle wave and to facilitate "hysteresis mode" switching.
See Application Note AN9325 for HIP4081, document \#9325. Intersil web home page: http://www.intersil.com
Similar part HIP4081A includes undervoltage circuitry which does not require the circuitry shown in Figure 30 of this data sheet.

## Pinout

HIP4081
(PDIP, SOIC)
TOP VIEW


## Application Block Diagram



Functional Block Diagram (1/2 HIP4081)


## Typical Application (PWM Mode Switching)



## Absolute Maximum Ratings

 Logic I/O Voltages ................................. 0.3 V to $\mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V}$ Voltage on AHS, BHS . . . -6.0V (Transient) to $80 \mathrm{~V}\left(25^{\circ} \mathrm{C}\right.$ to $\left.125^{\circ} \mathrm{C}\right)$ Voltage on AHS, BHS . . . -6.0 V (Transient) to $70 \mathrm{~V}\left(-55^{\circ} \mathrm{C}\right.$ to $\left.125^{\circ} \mathrm{C}\right)$ Voltage on ALS, BLS . . . . . . . -2.0V (Transient) to +2.0V (Transient) Voltage on AHB, BHB....... $\mathrm{V}_{\mathrm{AHS}}$, BHS -0.3 V to $\mathrm{V}_{\mathrm{AHS}}$, $\mathrm{BHS}+16 \mathrm{~V}$ Voltage on ALO, BLO ............. $\mathrm{V}_{\text {ALS }}$, BLS -0.3 V to $\mathrm{V}_{\mathrm{CC}}+0.3 \mathrm{~V}$ Voltage on AHO, BHO $\ldots . . V_{\text {AHS }}$, BHS -0.3 V to $\mathrm{V}_{\mathrm{AHB}}$, BHB +0.3 V Input Current, HDEL and LDEL . . . . . . . . . . . . . . . . . . . . 5 mA to 0 mA Phase Slew Rate . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 20V/ns
NOTE: All voltages are relative to pin $4, \mathrm{~V}_{\mathrm{SS}}$, unless otherwise specified.

Thermal Information (Typical, Note 1)
Storage Temperature Range . . . . . . . . . . . . . . . . . . . $65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$
Operating Max. Junction Temperature . . . . . . . . . . . . . . . . . . . $125^{\circ} \mathrm{C}$
Lead Temperature (Soldering 10s) . . . . . . . . . . . . . . . . . . . . . $300^{\circ} \mathrm{C}$
(For SOIC - Lead Tips Only)
Thermal Resistance, Junction-Ambient
SOIC Package . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . $85^{\circ} \mathrm{C} / \mathrm{W}$
DIP Package . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . $75^{\circ} \mathrm{C} / \mathrm{W}$

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTE:

1. $\theta_{\mathrm{JA}}$ is measured with the component mounted on an evaluation PC board in free air.

## Operating Conditions

Supply Voltage, $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{CC}} \ldots \ldots . . . . . . . . .$.
Voltage on ALS, BLS
-1.0 V to +1.0 V

Voltage on AHB, BHB $\ldots \ldots . V_{\text {AHS }}$, BHS +5 V to $\mathrm{V}_{\text {AHS }}$, BHS +15 V Input Current, HDEL and LDEL. . . . . . . . . . . . . . . . $-500 \mu \mathrm{~A}$ to $-50 \mu \mathrm{~A}$ Operating Ambient Temperature Range $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$

Electrical Specifications $\quad V_{D D}=V_{C C}=V_{A H B}=V_{B H B}=12 \mathrm{~V}, V_{S S}=V_{A L S}=V_{B L S}=V_{A H S}=V_{B H S}=0 V, R_{H D E L}=R_{L D E L}=100 \mathrm{~K}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, Unless Otherwise Specified

| PARAMETER | SYMBOL | TEST CONDITIONS | $\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C}$ |  |  | $\begin{gathered} \mathrm{T}_{\mathrm{JS}}=-40^{\circ} \mathrm{C} \\ \mathrm{TO} 125^{\circ} \mathrm{C} \end{gathered}$ |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | MAX |  |

SUPPLY CURRENTS AND CHARGE PUMPS

| $\mathrm{V}_{\mathrm{DD}}$ Quiescent Current | IDD | All Inputs = 0V | 7 | 9 | 11 | 6 | 12 | mA |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}$ Operating Current | IDDO | Outputs Switching $\mathrm{f}=500 \mathrm{kHz}$ | 8 | 9.5 | 12 | 7 | 13 | mA |
| $\mathrm{V}_{\text {CC }}$ Quiescent Current | $\mathrm{I}_{\mathrm{cc}}$ | All Inputs $=0 \mathrm{~V}, \mathrm{I}_{\text {ALO }}=\mathrm{I}_{\mathrm{BLO}}=0$ | - | 0.1 | 10 | - | 20 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {CC }}$ Operating Current | ICco | $f=500 \mathrm{kHz}$, No Load | 1 | 1.25 | 2.0 | 0.8 | 3 | mA |
| AHB, BHB Quiescent Current Qpump Output Current | ${ }^{\text {IAHB }}$, $\mathrm{I}_{\text {BHB }}$ | $\begin{aligned} & \text { All Inputs }=0 \mathrm{~V}, \mathrm{I}_{\mathrm{AHO}}=\mathrm{I}_{\mathrm{BHO}}=0 \\ & \mathrm{~V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{AHB}}=\mathrm{V}_{\mathrm{BHB}}=10 \mathrm{~V} \end{aligned}$ | -50 | -30 | -15 | -60 | -10 | $\mu \mathrm{A}$ |
| AHB, BHB Operating Current | ${ }^{\text {AHBO}}{ }^{\text {I }}$ I ${ }_{\text {BHBO }}$ | $\mathrm{f}=500 \mathrm{kHz}$, No Load | 0.5 | 0.9 | 1.3 | 0.4 | 1.7 | mA |
| AHS, BHS, AHB, BHB Leakage Current | IHLK | $\mathrm{V}_{\text {AHS }}=\mathrm{V}_{\mathrm{BHS}}=\mathrm{V}_{\mathrm{AHB}}=\mathrm{V}_{\mathrm{BHB}}=95 \mathrm{~V}$ | - | 0.02 | 1.0 | - | 10 | $\mu \mathrm{A}$ |
| AHB-AHS, BHB-BHS Qpump Output Voltage | $\begin{aligned} & \mathrm{V}_{\mathrm{AHB}}-\mathrm{V}_{\mathrm{AHS}} \\ & \mathrm{~V}_{\mathrm{BHB}}-\mathrm{V}_{\mathrm{BHS}} \end{aligned}$ | ${ }^{\text {I }}$ AB $=\mathrm{I}_{\text {AHB }}=0$, No Load | 11.5 | 12.6 | 14.0 | 10.5 | 14.5 | V |

INPUT PINS: ALI, BLI, AHI, BHI, AND DIS

| Low Level Input Voltage | $\mathrm{V}_{\mathrm{IL}}$ | Full Operating Conditions | - | - | 1.0 | - | 0.8 | V |
| :--- | :---: | :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| High Level Input Voltage | $\mathrm{V}_{\mathrm{IH}}$ | Full Operating Conditions | 2.5 | - | - | 2.7 | - | V |
| Input Voltage Hysteresis |  |  | - | 35 | - | - | - | mV |
| Low Level Input Current | $\mathrm{I}_{\mathrm{IL}}$ | $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$, Full Operating Conditions | -130 | -100 | -75 | -135 | -65 | $\mu \mathrm{~A}$ |
| High Level Input Current | $\mathrm{I}_{\mathrm{IH}}$ | $\mathrm{V}_{\mathrm{IN}}=5 \mathrm{~V}$, Full Operating Conditions | -1 | - | +1 | -10 | +10 | $\mu \mathrm{~A}$ |

TURN-ON DELAY PINS: LDEL AND HDEL

| LDEL, HDEL Voltage | $V_{\text {HDEL, }} V_{\text {LDEL }}$ | $I_{\text {HDEL }}=I_{\text {LDEL }}=-100 \mu \mathrm{~A}$ | 4.9 | 5.1 | 5.3 | 4.8 | 5.4 | V |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: |

GATE DRIVER OUTPUT PINS: ALO, BLO, AHO, AND BHO

| Low Level Output Voltage | $\mathrm{V}_{\mathrm{OL}}$ | $\mathrm{I}_{\text {OUT }}=100 \mathrm{~mA}$ | 0.7 | 0.85 | 1.0 | 0.5 | 1.1 | V |
| :--- | :---: | :--- | :--- | :--- | :--- | :--- | :--- | :---: |
| High Level Output Voltage | $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{OH}}$ | $\mathrm{I}_{\mathrm{OUT}}=-100 \mathrm{~mA}$ | 0.8 | .95 | 1.1 | 0.5 | 1.2 | V |

HIP4081

Electrical Specifications $\quad V_{D D}=V_{C C}=V_{A H B}=V_{B H B}=12 \mathrm{~V}, V_{S S}=V_{A L S}=V_{B L S}=V_{A H S}=V_{B H S}=0 V, R_{H D E L}=R_{L D E L}=100 \mathrm{~K}$ and $T_{A}=25^{\circ} \mathrm{C}$, Unless Otherwise Specified (Continued)

| PARAMETER | SYMBOL | TEST CONDITIONS | $\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C}$ |  |  | $\begin{gathered} \mathrm{T}_{\mathrm{JS}}=-40^{\circ} \mathrm{C} \\ \mathrm{TO} 125^{\circ} \mathrm{C} \end{gathered}$ |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | MAX |  |
| Peak Pullup Current | $\mathrm{I}^{+}$ | $\mathrm{V}_{\text {OUT }}=0 \mathrm{~V}$ | 1.7 | 2.6 | 3.8 | 1.4 | 4.1 | A |
| Peak Pulldown Current | $10^{-}$ | $\mathrm{V}_{\text {OUT }}=12 \mathrm{~V}$ | 1.7 | 2.4 | 3.3 | 1.3 | 3.6 | A |

Switching Specifications $\quad V_{D D}=V_{C C}=V_{A H B}=V_{B H B}=12 V, V_{S S}=V_{A L S}=V_{B L S}=V_{A H S}=V_{B H S}=0 V, R_{H D E L}=R_{L D E L}=10 K$, $C_{L}=1000 \mathrm{pF}$

| PARAMETER | SYMBOL | TEST CONDITIONS | $\mathrm{T}_{\mathrm{J}}=+25^{\circ} \mathrm{C}$ |  |  | $\begin{aligned} & \hline \mathrm{T}_{\mathrm{JS}}=40^{\circ} \mathrm{C} \\ & \text { TO } 125^{\circ} \mathrm{C} \end{aligned}$ |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | MAX |  |
| Lower Turn-off Propagation Delay (ALI-ALO, BLI-BLO) | TLPHL |  | - | 30 | 60 | - | 80 | ns |
| Upper Turn-off Propagation Delay (AHI-AHO, BHI-BHO) | $\mathrm{T}_{\text {HPHL }}$ |  | - | 35 | 70 | - | 90 | ns |
| Lower Turn-on Propagation Delay (ALI-ALO, BLI-BLO) | TLPLH |  | - | 45 | 70 | - | 90 | ns |
| Upper Turn-on Propagation Delay (AHI-AHO, BHI-BHO) | $\mathrm{T}_{\text {HPLH }}$ |  | - | 60 | 90 | - | 110 | ns |
| Rise Time | $\mathrm{T}_{\mathrm{R}}$ |  | - | 10 | 25 | - | 35 | ns |
| Fall Time | $\mathrm{T}_{\mathrm{F}}$ |  | - | 10 | 25 | - | 35 | ns |
| Turn-on Input Pulse Width | TPWIN-ON |  | 50 | - | - | 50 | - | ns |
| Turn-off Input Pulse Width | TPWIN-OFF |  | 40 | - | - | 40 | - | ns |
| Disable Turn-off Propagation Delay (DIS - Lower Outputs) | T DISLOW |  | - | 45 | 75 | - | 95 | ns |
| Disable Turn-off Propagation Delay (DIS - Upper Outputs) | T ${ }_{\text {DISHIGH }}$ |  | - | 55 | 85 | - | 105 | ns |
| Disable to Lower Turn-on Propagation Delay (DIS - ALO and BLO) | T DLPLH |  | - | 35 | 70 | - | 90 | ns |
| Refresh Pulse Width (ALO and BLO) | TREF-PW |  | 160 | 260 | 380 | 140 | 420 | ns |
| Disable to Upper Enable (DIS - AHO and BHO) | THEN |  | - | 335 | 500 | - | 550 | ns |

TRUTH TABLE

| INPUT |  | OUTPUT |  |  |
| :---: | :---: | :---: | :---: | :---: |
| ALI, BLI | AHI, BHI | DIS | ALO, BLO | AHO, BHO |
| $X$ | $X$ | 1 | 0 | 0 |
| 1 | $X$ | 0 | 1 | 0 |
| 0 | 1 | 0 | 0 | 1 |
| 0 | 0 | 0 | 0 | 0 |

NOTE: X signifies that input can be either a " 1 " or " 0 ".

## Pin Descriptions

| PIN NUMBER | SYMBOL | DESCRIPTION |
| :---: | :---: | :---: |
| 1 | BHB | B High-side Bootstrap supply. External bootstrap diode and capacitor are required. Connect cathode of bootstrap diode and positive side of bootstrap capacitor to this pin. Internal charge pump supplies $30 \mu \mathrm{~A}$ out of this pin to maintain bootstrap supply. Internal circuitry clamps the bootstrap supply to approximately 12.8 V . |
| 2 | BHI | B High-side Input. Logic level input that controls BHO driver (Pin 20). BLI (Pin 5) high level input overrides BHI high level input to prevent half-bridge shoot-through, see Truth Table. DIS (Pin 3) high level input overrides BHI high level input. The pin can be driven by signal levels of 0 V to 15 V (no greater than $\mathrm{V}_{\mathrm{DD}}$ ). An internal $100 \mu \mathrm{~A}$ pull-up to $\mathrm{V}_{\mathrm{DD}}$ will hold BHI high, so no connection is required if high-side and low-side outputs are to be controlled by the low-side input. |
| 3 | DIS | Disable input. Logic level input that when taken high sets all four outputs low. DIS high overrides all other inputs. When DIS is taken low the outputs are controlled by the other inputs. The pin can be driven by signal levels of 0 V to 15 V (no greater than $\mathrm{V}_{\mathrm{DD}}$ ). An internal $100 \mu \mathrm{~A}$ pull-up to $\mathrm{V}_{\mathrm{DD}}$ will hold DIS high if this pin is not driven. |
| 4 | $\mathrm{V}_{\text {SS }}$ | Chip negative supply, generally will be ground. |
| 5 | BLI | B Low-side Input. Logic level input that controls BLO driver (Pin 18). If BHI (Pin 2) is driven high or not connected externally then BLI controls both BLO and BHO drivers, with dead time set by delay currents at HDEL and LDEL (Pin 8 and 9). DIS (Pin 3) high level input overrides BLI high level input. The pin can be driven by signal levels of 0 V to 15 V (no greater than $\mathrm{V}_{\mathrm{DD}}$ ). An internal $100 \mu \mathrm{~A}$ pull-up to $\mathrm{V}_{\mathrm{DD}}$ will hold BLI high if this pin is not driven. |
| 6 | ALI | A Low-side Input. Logic level input that controls ALO driver (Pin 13). If AHI (Pin 7) is driven high or not connected externally then ALI controls both ALO and AHO drivers, with dead time set by delay currents at HDEL and LDEL (Pin 8 and 9). DIS (Pin 3) high level input overrides ALI high level input. The pin can be driven by signal levels of 0 V to 15 V (no greater than $\mathrm{V}_{\mathrm{DD}}$ ). An internal $100 \mu \mathrm{~A}$ pull-up to $\mathrm{V}_{\mathrm{DD}}$ will hold ALI high if this pin is not driven. |
| 7 | AHI | A High-side Input. Logic level input that controls AHO driver (Pin 11). ALI (Pin 6) high level input overrides AHI high level input to prevent half-bridge shoot-through, see Truth Table. DIS (Pin 3) high level input overrides AHI high level input. The pin can be driven by signal levels of 0 V to 15 V (no greater than $\mathrm{V}_{\mathrm{DD}}$ ). An internal $100 \mu \mathrm{~A}$ pull-up to $\mathrm{V}_{\mathrm{DD}}$ will hold AHI high, so no connection is required if high-side and low-side outputs are to be controlled by the low-side input. |
| 8 | HDEL | High-side turn-on DELay. Connect resistor from this pin to $\mathrm{V}_{\mathrm{SS}}$ to set timing current that defines the turn-on delay of both high-side drivers. The low-side drivers turn-off with no adjustable delay, so the HDEL resistor guarantees no shoot-through by delaying the turn-on of the high-side drivers. HDEL reference voltage is approximately 5.1 V . |
| 9 | LDEL | Low-side turn-on DELay. Connect resistor from this pin to $\mathrm{V}_{\text {SS }}$ to set timing current that defines the turn-on delay of both low-side drivers. The high-side drivers turn-off with no adjustable delay, so the LDEL resistor guarantees no shoot-through by delaying the turn-on of the low-side drivers. LDEL reference voltage is approximately 5.1 V . |
| 10 | AHB | A High-side Bootstrap supply. External bootstrap diode and capacitor are required. Connect cathode of bootstrap diode and positive side of bootstrap capacitor to this pin. Internal charge pump supplies $30 \mu \mathrm{~A}$ out of this pin to maintain bootstrap supply. Internal circuitry clamps the bootstrap supply to approximately 12.8 V . |
| 11 | AHO | A High-side Output. Connect to gate of A High-side power MOSFET. |
| 12 | AHS | A High-side Source connection. Connect to source of A High-side power MOSFET. Connect negative side of bootstrap capacitor to this pin. |
| 13 | ALO | A Low-side Output. Connect to gate of A Low-side power MOSFET. |
| 14 | ALS | A Low-side Source connection. Connect to source of A Low-side power MOSFET. |
| 15 | $\mathrm{V}_{\mathrm{CC}}$ | Positive supply to gate drivers. Must be same potential as $\mathrm{V}_{\mathrm{DD}}$ (Pin 16). Connect to anodes of two bootstrap diodes. |
| 16 | $\mathrm{V}_{\mathrm{DD}}$ | Positive supply to lower gate drivers. Must be same potential as $\mathrm{V}_{\text {CC }}$ (Pin 15). De-couple this pin to $\mathrm{V}_{\text {SS }}$ (Pin 4). |
| 17 | BLS | B Low-side Source connection. Connect to source of B Low-side power MOSFET. |
| 18 | BLO | B Low-side Output. Connect to gate of B Low-side power MOSFET. |
| 19 | BHS | B High-side Source connection. Connect to source of B High-side power MOSFET. Connect negative side of bootstrap capacitor to this pin. |
| 20 | BHO | B High-side Output. Connect to gate of B High-side power MOSFET. |

## Timing Diagrams



FIGURE 1. INDEPENDENT MODE


FIGURE 2. BISTATE MODE


FIGURE 3. DISABLE FUNCTION

Typical Performance Curves $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{AHB}}=\mathrm{V}_{\mathrm{BHB}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=\mathrm{V}_{\mathrm{ALS}}=\mathrm{V}_{\mathrm{BLS}}=\mathrm{V}_{\mathrm{AHS}}=\mathrm{V}_{\mathrm{BHS}}=0 \mathrm{~V}$, $R_{\text {HDEL }}=R_{\text {LDEL }}=100 \mathrm{~K}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, Unless Otherwise Specified


FIGURE 4. QUIESCENT IDD SUPPLY CURRENT vs $V_{D D}$ SUPPLY VOLTAGE


FIGURE 6. SIDE A, B FLOATING SUPPLY BIAS CURRENT vs FREQUENCY (LOAD = 1000pF)


FIGURE 8. $\mathrm{I}_{\mathrm{AHB}}, \mathrm{I}_{\mathrm{BHB}}$ NO-LOAD FLOATING SUPPLY CURRENT vs FREQUENCE


FIGURE 5. IDDO, NO-LOAD IDD SUPPLY CURRENT vs FREQUENCY (kHz)


FIGURE 7. Icco, NO-LOAD Icc SUPPLY CURRENT vs FREQUENCY (kHz) TEMPERATURE


FIGURE 9. ALI, BLI, AHI, BHI LOW LEVEL INPUT CURRENT IIL vs TEMPERATURE

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Typical Performance Curves $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{AHB}}=\mathrm{V}_{\mathrm{BHB}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=\mathrm{V}_{\mathrm{ALS}}=\mathrm{V}_{\mathrm{BLS}}=\mathrm{V}_{\mathrm{AHS}}=\mathrm{V}_{\mathrm{BHS}}=0 \mathrm{~V}$, $R_{\text {HDEL }}=R_{\text {LDEL }}=100 \mathrm{~K}$ and $\mathrm{T}_{A}=25^{\circ} \mathrm{C}$, Unless Otherwise Specified (Continued)


FIGURE 10. AHB - AHS, BHB - BHS NO-LOAD CHARGE PUMP VOLTAGE vs TEMPERATURE


FIGURE 12. DISABLE TO UPPER ENABLE TUEN PROPAGATION DELAY vs TEMPERATURE


FIGURE 14. TREF-PW REFRESH PULSE WIDTH vs TEMPERATURE


FIGURE 11. UPPER DISABLE TURN-OFF PROPAGATION DELAY TDISHIGH vs TEMPERATURE


FIGURE 13. DISABLE TO UPPER ENABLE TUEN PROPAGATION DELAY vs TEMPERATURE


FIGURE 15. DISABLE TO LOWER ENABLE TDLPLH PROPAGATION DELAY vs TEMPERATURE

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FIGURE 16. UPPER TURN-OFF PROPAGATION DELAY THPHL vs TEMPERATURE


FIGURE 18. LOWER TURN-OFF PROPAGATION DELAY TLPHL vs TEMPERATURE


FIGURE 20. GATE DRIVE FALL TIME $T_{F}$ vs TEMPERATURE


FIGURE 17. UPPER TURN-ON PROPAGATION DELAY THPLH $v s$ TEMPERATURE


FIGURE 19. LOWER TURN-ON PROPAGATION DELAY TLPLH TEMPERATURE


FIGURE 21. GATE DRIVE RISE TIME $T_{R}$ vs TEMPERATURE

Typical Performance Curves $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{AHB}}=\mathrm{V}_{\mathrm{BHB}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=\mathrm{V}_{\mathrm{ALS}}=\mathrm{V}_{\mathrm{BLS}}=\mathrm{V}_{\mathrm{AHS}}=\mathrm{V}_{\mathrm{BHS}}=0 \mathrm{~V}$, $R_{\text {HDEL }}=R_{\text {LDEL }}=100 \mathrm{~K}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, Unless Otherwise Specified (Continued)


FIGURE 22. V LDeL , $\mathrm{V}_{\text {hdel }}$ VOLTAGE vs TEMPERATURE


FIGURE 24. LOW LEVEL OUTPUT VOLTAGE VOL vs BIAS SUPPLY AND TEMPERTURE AT 100mA


FIGURE 26. PEAK PULLUP CURRENT IO+ vs BIAS SUPPLY VOLTAGE


FIGURE 23. HIGH LEVEL OUTPUT VOLTAGE VCC ${ }_{\text {- }}^{\text {OH }}$ vs BIAS SUPPLY AND TEMPERATURE AT 100mA


FIGURE 25. PEAK PULLDOWN CURRENT Io vs BIAS SUPPLY VOLTAGE


FIGURE 27. LOW VOLTAGE BIAS CURRENT IDD (LESS QUIESCENT COMPONENT) vs FREQUENCY AND GATE LOAD CAPACITANCE
$\begin{aligned} & \text { Typical Performance Curves } \mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{C C}=\mathrm{V}_{A H B}=\mathrm{V}_{B H B}=12 \mathrm{~V}, \mathrm{~V}_{S S}=\mathrm{V}_{A L S}=\mathrm{V}_{B L S}=\mathrm{V}_{A H S}=\mathrm{V}_{B H S}=0 \mathrm{~V}, \\ & R_{H D E L}=R_{L D E L}=100 \mathrm{~K} \text { and } \mathrm{T}_{A}=25^{\circ} \mathrm{C} \text {, Unless Otherwise Specified (Continued) }\end{aligned}$


FIGURE 28. HIGH VOLTAGE LEVEL-SHIFT CURRENT vs FREQUENCY AND BUS VOLTAGE


FIGURE 29. MINIMUM DEAD-TIME vs DEL RESISTANCE

## HI4081 Power-up Application Information

The HIP4081 H-Bridge Driver IC requires external circuitry to assure reliable start-up conditions of the upper drivers. If not addressed in the application, the H-bridge power MOSFETs may be exposed to shoot-through current, possibly leading to MOSFET failure. Following the instructions below will result in reliable start-up.
The HIP4081 has four inputs, one for each output. Outputs ALO and BLO are directly controlled by input ALI and BLI. By holding ALI and BLI low during start-up no shoot-through conditions can occur. To set the latches to the upper drivers such that the driver outputs, AHO and BHO, are off, the DIS pin must be toggled from low to high after power is applied. This is accomplished with a simple resistor divider, as shown below in Figure 30. As the $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{CC}}$ supply ramps from zero up, the DIS voltage is below its input threshold of 1.7 V due to the $\mathrm{R} 1 / \mathrm{R} 2$ resistor divider. When $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{CC}}$ exceeds approximately 9 V to 10 V , DIS becomes greater than the input threshold and the chip disables all outputs. It is critical that ALI and BLI be held low prior to DIS reaching its threshold level of 1.7 V while $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{CC}}$ is ramping up, so that shoot through is avoided. After power is up the chip can be enabled by the ENABLE signal which pulls the DIS pin low.


FIGURE 30A.


NOTES:
2. ALI and/or BLI may be high after t1, whereupon the ENABLE pin may also be brought high.
3. Another product, HIP4081A, incorporates undervoltage circuitry which eliminates the need for the above power up circuitry.

FIGURE 30B. TIMING DIAGRAM FOR FIGURE 30A


## Supplemental Information for HIP4080 and HIP4081 Power Application

The HIP4080 and HIP4081 H-Bridge Driver ICs require external circuitry to assure reliable start-up conditions of the upper drivers. If not addressed in the application, the H-bridge power MOSFETs may be exposed to shootthrough current, possibly leading to MOSFET failure. Following the instructions below will result in reliable start-up.

## HIP4081

The HIP4081 has four inputs, one for each output. Outputs ALO and BLO are directly controlled by input ALI and BLI. By holding ALI and BLI low during start-up no shoot-through conditions can occur. To set the latches to the upper drivers such that the driver outputs, AHO and BHO, are off, the DIS pin must be toggled from low to high after power is applied. This is accomplished with a simple resistor divider, as shown below in Figure 33. As the $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{CC}}$ supply ramps from zero up, the DIS voltage is below its input threshold of 1.7 V due to the $\mathrm{R} 1 / \mathrm{R} 2$ resistor divider. When $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{CC}}$ exceeds approximately 9 V to 10 V , DIS becomes greater than the input threshold and the chip disables all outputs. It is critical that ALI and BLI be held low prior to DIS reaching its threshold
level of 1.7 V while $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{CC}}$ is ramping up, so that shoot through is avoided. After power is up the chip can be enabled by the ENABLE signal which pulls the DIS pin low.

## HIP4080

The HIP4080 does not have an input protocol like the HIP4081 that keeps both lower power MOSFETs off other than through the DIS pin. IN+ and IN- are inputs to a comparator that control the bridge in such a way that only one of the lower power devices is on at a time, assuming DIS is low. However, keeping both lower MOSFETs off can be accomplished by controlling the lower turn-on delay pin, LDEL, while the chip is enabled, as shown in Figure 34. Pulling LDEL to $\mathrm{V}_{\text {DD }}$ will indefinitely delay the lower turn-on delays through the input comparator and will keep the lower MOSFETs off. With the lower MOSFETs off and the chip enabled, i.e., DIS = low, $\mathrm{IN}+$ or IN - can be switched through a full cycle, properly setting the upper driver outputs. Once this is accomplished, LDEL is released to its normal operating point. It is critical that $\mathrm{IN}+/ \mathrm{IN}$ - switch a full cycle while LDEL is held high, to avoid shoot-through. This start-up procedure can be initiated by the supply voltage and/or the chip enable command by the circuit in Figure 33.


FIGURE 33.


FIGURE 34.

## Timing Diagrams


6. ALI and/or BLI may be high after t1, whereupon the ENABLE pin may also be brought high.


NOTE:
7. Between $t 1$ and $t 2$ the $\mathrm{IN}+$ and IN - inputs must cause the OUT pin to go through one complete cycle (transition order is not important). If the ENABLE pin is low after the undervoltage circuit is satisfied, the ENABLE pin will initiate the 10ms time delay during which the $\mathrm{IN}+$ and IN - pins must cycle at least once.

FIGURE 36.

## Dual-In-Line Plastic Packages (PDIP)



NOTES:

1. Controlling Dimensions: $\operatorname{INCH}$. In case of conflict between English and Metric dimensions, the inch dimensions control.
2. Dimensioning and tolerancing per ANSI Y14.5M-1982.
3. Symbols are defined in the "MO Series Symbol List" in Section 2.2 of Publication No. 95.
4. Dimensions A, A1 and L are measured with the package seated in JEDEC seating plane gauge GS-3.
5. D, D1, and E1 dimensions do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.010 inch ( 0.25 mm ).
6. $E$ and $e_{A}$ are measured with the leads constrained to be perpendicular to datum $-\mathrm{C}-$.
7. $e_{B}$ and $e_{C}$ are measured at the lead tips with the leads unconstrained. ec must be zero or greater.
8. B1 maximum dimensions do not include dambar protrusions. Dambar protrusions shall not exceed 0.010 inch ( 0.25 mm ).
9. N is the maximum number of terminal positions.
10. Corner leads ( $1, \mathrm{~N}, \mathrm{~N} / 2$ and $\mathrm{N} / 2+1$ ) for E8.3, E16.3, E18.3, E28.3, E42.6 will have a B1 dimension of $0.030-0.045$ inch (0.76-1.14mm).

E20.3 (JEDEC MS-001-AD ISSUE D) 20 LEAD DUAL-IN-LINE PLASTIC PACKAGE

| SYMBOL | INCHES |  | MILLIMETERS |  | NOTES |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN | MAX | MIN | MAX |  |
| A | - | 0.210 | - | 5.33 | 4 |
| A1 | 0.015 | - | 0.39 | - | 4 |
| A2 | 0.115 | 0.195 | 2.93 | 4.95 | - |
| B | 0.014 | 0.022 | 0.356 | 0.558 | - |
| B1 | 0.045 | 0.070 | 1.55 | 1.77 | 8 |
| C | 0.008 | 0.014 | 0.204 | 0.355 | - |
| D | 0.980 | 1.060 | 24.89 | 26.9 | 5 |
| D1 | 0.005 | - | 0.13 | - | 5 |
| E | 0.300 | 0.325 | 7.62 | 8.25 | 6 |
| E1 | 0.240 | 0.280 | 6.10 | 7.11 | 5 |
| e |  | BC |  | BSC | - |
| $\mathrm{e}_{\mathrm{A}}$ |  | BC |  | BSC | 6 |
| $\mathrm{e}_{\mathrm{B}}$ | - | 0.430 | - | 10.92 | 7 |
| L | 0.115 | 0.150 | 2.93 | 3.81 | 4 |
| N | 20 |  | 20 |  | 9 |

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## Small Outline Plastic Packages (SOIC)



1. Symbols are defined in the "MO Series Symbol List" in Section 2.2 of Publication Number 95.
2. Dimensioning and tolerancing per ANSI Y14.5M-1982.
3. Dimension " $D$ " does not include mold flash, protrusions or gate burrs. Mold flash, protrusion and gate burrs shall not exceed 0.15 mm ( 0.006 inch) per side.
4. Dimension " $E$ " does not include interlead flash or protrusions. Interlead flash and protrusions shall not exceed 0.25 mm ( 0.010 inch) per side.
5. The chamfer on the body is optional. If it is not present, a visual index feature must be located within the crosshatched area.
6. " $L$ " is the length of terminal for soldering to a substrate.
7. " N " is the number of terminal positions.
8. Terminal numbers are shown for reference only.
9. The lead width " $B$ ", as measured 0.36 mm ( 0.014 inch) or greater above the seating plane, shall not exceed a maximum value of 0.61 mm ( 0.024 inch )
10. Controlling dimension: MILLIMETER. Converted inch dimensions are not necessarily exact.

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